#### **Features**

- Wide Input Voltage Range 4.5V to 52V
- $52V / 0.9\Omega$  Internal Power MOSFET
- 600mA Peak Output Current
- Up to 90% Efficiency
- 1.25MHz (HT7463A) and 550kHz (HT7463B) Fixed Operating Frequency
- Ultra Low Shutdown Current  $< 1\mu A$
- · Output Short Circuit Protection
- Thermal Shutdown Protection
- Package Type: 6-pin SOT23

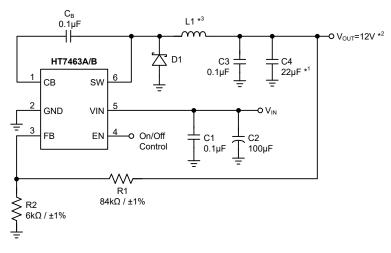
## **Applications**

- · Power Meters
- · Distribution Power Systems
- · Battery Chargers
- Pre-Regulator for Linear Regulators

## **General Description**

The HT7463A/B is a current mode buck converter. With a wide input range from 4.5V to 52V, the HT7463A/B is suitable for a wide range of applications such as power conditioning from unregulated sources. Having a low internal switch typical RDSON value of  $0.9\Omega$ , the device has a good operating typical efficiency value of 85% and the added advantage of reduced junction temperature. The operating frequency is fixed at 1250/550kHz for the HT7463A/HT7463B respectively. The HT7463A allows the use of small external components while still being able to have low output voltage ripple. A soft-start function can be implemented using the enable pin and by connecting an external RC circuit allowing the user to tailor the soft-start time to a specific application.

## **Application Circuit**



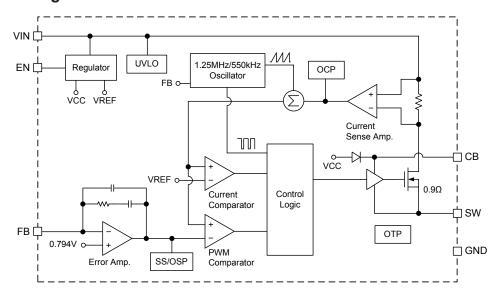
Note: \*1. C4=330 $\mu F$  is recommended to achieve 1‰ output ripple requirement

- \*2. Set R1=84k $\Omega$  and R2=6k $\Omega$  for  $V_{\text{OUT}}$ =12V application
- \*3. Typically recommended that L1=22μH for HT7463A and L1=47μH for HT7463B. Electromagnetic interference situation suggest L1=100μH or more

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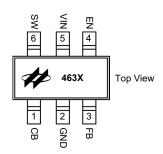


# **Block Diagram**



# **Pin Assignment**

#### SOT23-6



X means A(1.25MHz)/or B(550kHz)

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# **Pin Descpription**

Pin Order	Name	Туре	Pin Discpription	
1	СВ	I/O	SW FET gate bias voltage. Connect the boot capacitor between CB and SW	
2	GND	G	Ground terminal	
3	FB	I	eedback pin: Set feedback voltage divider ratio with V <sub>OUT</sub> = V <sub>FB</sub> (1+(R1/R2))	
4	EN	I	ogic level shutdown pin. Internal pull low resistor	
5	VIN	Р	Power supply	
6	SW	0	Power FET output	

# **Absolute Maximum Ratings**

	Parameter	Value	Unit
VIN and SW		-0.3 to +55	V
EN		-0.3 to (V <sub>IN</sub> +0.3)	V
CB above SW voltage		+5.5	V
FB		-0.3 to +5.0	V
Operating Temperature Range		-40 to +85	°C
Maximum Junction Temperature		+150	°C
Storage Temperature Range		-65 to +160	°C
Lead Temperature (Soldering 10sec)		+300	°C
CCD Cuscontibility	Human Body Model	2000	V
ESD Susceptibility	Machine Model	200	V
Junction-to-Ambient Thermal Resistance, θ <sub>JA</sub>		220	°C/W
Junction-to-Case Thermal Resistance, θ <sub>JC</sub>		110	°C/W

# **Recommended Operating Range**

Parameter	Value	Unit
VIN	4.5 to 52	V
SW and EN	Up to 52	V

Note that Absolute Maximum Ratings indicate limitations beyond which damage to the device may occur. Recommended Operating Ratings indicate conditions for which the device is intended to be functional, but do not guarantee specified performance limits.

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# **Electrical Characteristics**

 $V_{\text{IN}}$ =12V and Ta=+25°C, unless otherwise specified

Symbol	Parameter	Test Condition	Min	Тур	Max	Unit	
Supply Vo	Supply Voltage						
V <sub>IN</sub>	Input Voltage	VIN	4.5	_	52	V	
I <sub>cc</sub>	Quiescent Current	V <sub>EN</sub> =2.5V, V <sub>FB</sub> =1V	_	0.7	1	mA	
I <sub>OFF</sub>	Shutdown Current	V <sub>EN</sub> =0V	_	0.1	1	μΑ	
Buck Con	verter						
V <sub>OUT</sub>	Output Voltage (Note)	_	1.0	_	0.9×V <sub>IN</sub>	V	
	Cuitobing Fraguency	HT7463A, V <sub>FB</sub> =0.6V	1000	1250	1500	kHz	
f <sub>sw</sub>	Switching Frequency	HT7463B, V <sub>FB</sub> =0.6V	440	550	660	kHz	
_	Fold-back Frequency	HT7463A, V <sub>FB</sub> =0V	90	105	_	kHz	
F <sub>FB</sub>	Fold-back Frequency	HT7463B, V <sub>FB</sub> =0V	90	105	_	kHz	
	Maximum Duty Cycle	HT7463A	_	90	_	%	
D <sub>MAX</sub>	Maximum Duty Cycle	HT7463B	_	95	_	%	
$T_{ON(min)}$	Minimum ON-Time	_	_	100	_	ns	
R <sub>DS(on)</sub>	Switch-ON Resistance	V <sub>EN</sub> =2.5V	_	0.9	_	Ω	
I <sub>SW(off)</sub>	SW Leakage Current	$V_{EN}$ =0V, $V_{SW}$ =0V, $V_{IN}$ =52V	_	0.1	1	μA	
$V_{FB}$	Feedback Voltage	$4.5V \le V_{IN} \le 52V$	0.778	0.794	0.81	V	
I <sub>FB(leak)</sub>	Feedback Leakage Current	V <sub>FB</sub> =3V	_	_	0.1	μΑ	
	EN Input Current	V <sub>EN</sub> =0V	_	0.1	_	μΑ	
I <sub>EN</sub>	Liv input Guirent	V <sub>EN</sub> =52V	_	16	_	μΑ	
V <sub>IH</sub>	EN High Voltage Threshold	$4.5V \le V_{IN} \le 52V$	2.3		_	V	
V <sub>IL</sub>	EN Low Voltage Threshold	$4.5V \le V_{IN} \le 52V$	_	_	0.9	V	
Protections							
V <sub>UVLO+</sub>	Input Supply Turn ON Level	UVLO+		_	4.2	V	
V <sub>UVLO-</sub>	Input Supply Turn OFF Level	UVLO-	3.4	_	_	V	
I <sub>OCP</sub>	Over Current Protection Threshold			1	_	Α	
T <sub>SHD</sub>	Thermal Shutdown Threshold	OTP	_	150	_	°C	
T <sub>REC</sub>	Thermal Recovery Temperature	_	_	125	_	°C	

Note: 1. MIN Output Voltage is restricted by Minimum ON-Time, 100ns.

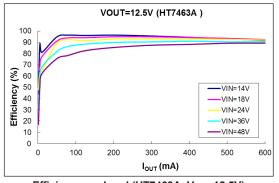
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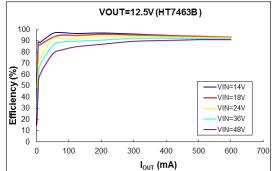
<sup>2.</sup> MAX Output Voltage is restricted by Maximum Duty Cycle and Switch-ON Resistance.



## **Typical Performance Characteristics**

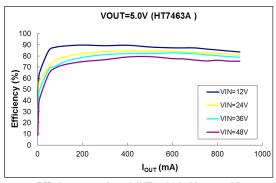
 $V_{\text{IN}}\text{=}18\text{V}, V_{\text{OUT}}\text{=}12.5\text{V}, L\text{=}15/22\mu\text{H for HT7463A and L=}33/47\mu\text{H for HT7463B}, TA=25^{\circ}\text{C}, unless otherwise noted to the state of th$ 

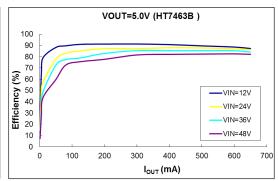




Efficiency vs. Load (HT7463A, V<sub>OUT</sub>=12.5V)

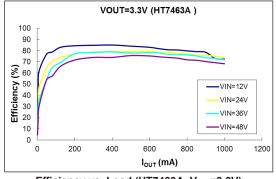
Efficiency vs. Load (HT7463B, V<sub>out</sub>=12.5V)

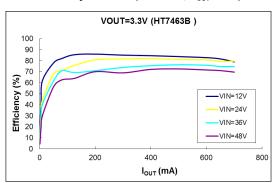




Efficiency vs. Load (HT7463A, V<sub>OUT</sub>=5.7V)

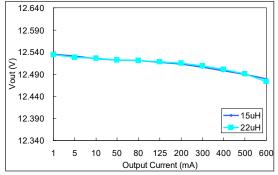
Efficiency vs. Load (HT7463B, V<sub>our</sub>=5.7V)

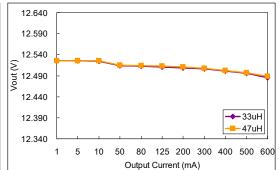




Efficiency vs. Load (HT7463A, V<sub>OUT</sub>=3.3V)

Efficiency vs. Load (HT7463B, V<sub>OUT</sub>=3.3V)





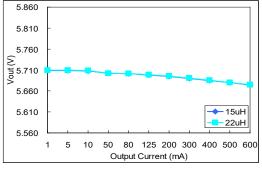
Load Regulation (HT7463A, V<sub>out</sub>=12.5V)

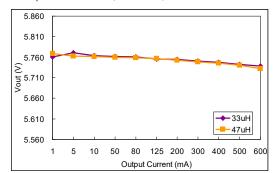
Load Regulation (HT7463B, V<sub>out</sub>=12.5V)



## **Typical Performance Characteristics (Continued)**

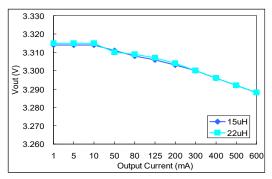
 $V_{\text{IN}}$ =18V,  $V_{\text{OUT}}$ =12.5V, L=22 $\mu$ H for HT7463A and L=47 $\mu$ H for HT7463B, TA=25 $^{\circ}$ C, unless otherwise noted

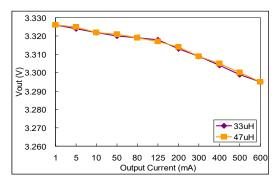




Load Regulation (HT7463A, V<sub>out</sub>=5.7V)

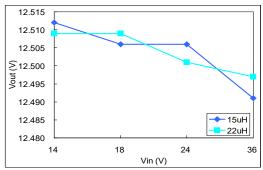
Load Regulation (HT7463B, V<sub>out</sub>=5.7V)

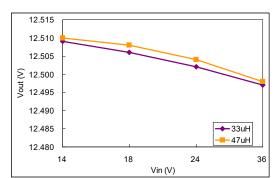




Load Regulation (HT7463A, V<sub>OUT</sub>=3.3V)

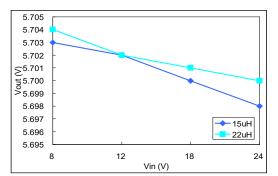
Load Regulation (HT7463B, V<sub>out</sub>=3.3V)

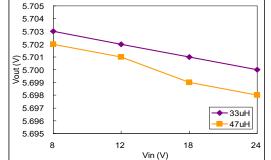




Line Regulation (HT7463A, V<sub>OUT</sub>=12.5V, I<sub>OUT</sub>=300mA)

Line Regulation (HT7463B, V<sub>OUT</sub>=12.5V, I<sub>OUT</sub>=300mA)

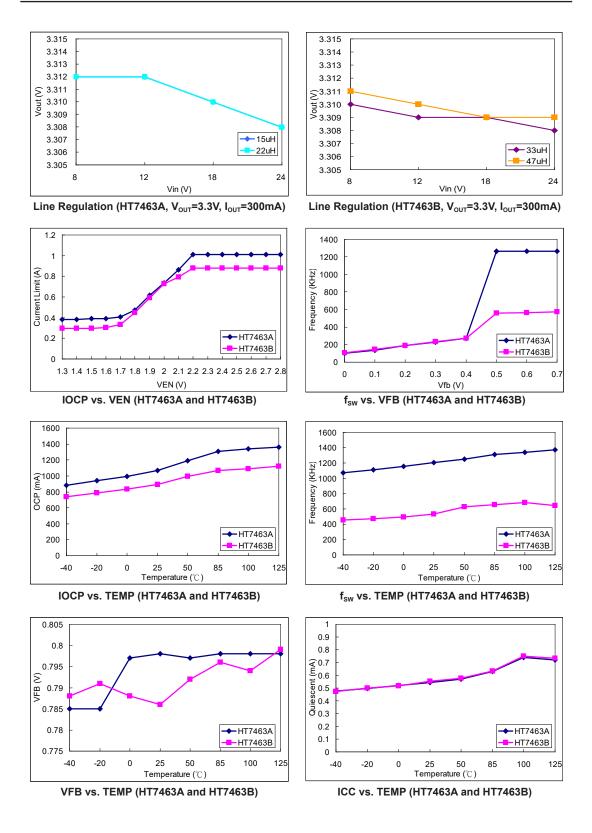




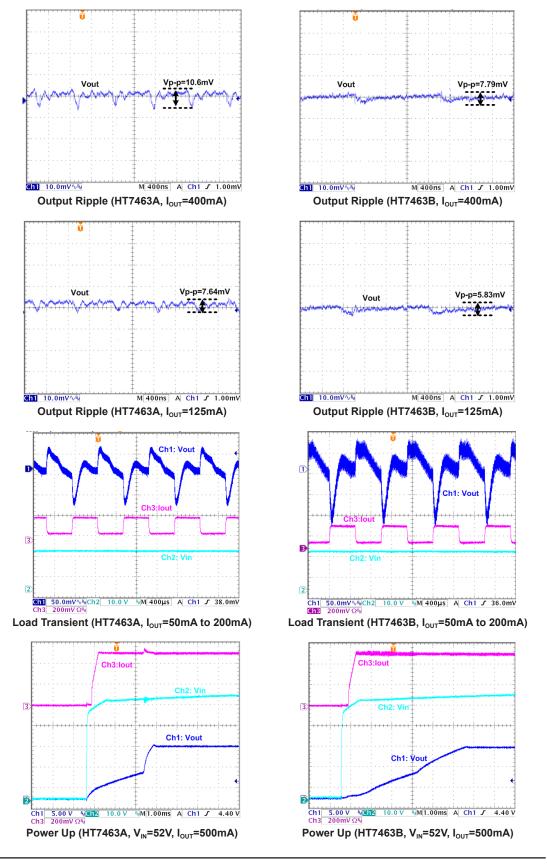
Line Regulation (HT7463A, V<sub>OUT</sub>=5.7V, I<sub>OUT</sub>=300mA)

Line Regulation (HT7463B, V<sub>OUT</sub>=5.7V, I<sub>OUT</sub>=300mA)

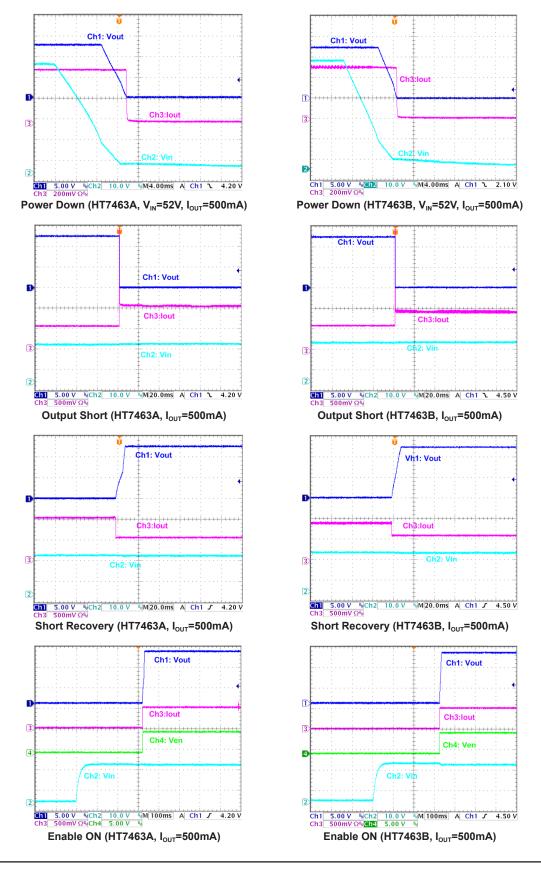




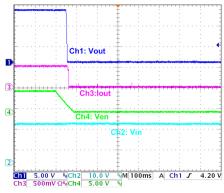




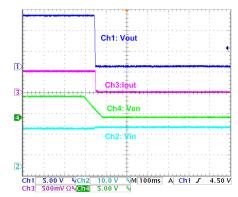












Enable OFF (HT7463B, I<sub>OUT</sub>=500mA)

# **Functional Description**

#### **Output Voltage Setup**

The external resistor divider sets the output voltage, for details see the Application Circuit. The feedback resistor, R1, also sets the feedback loop bandwidth with the internal compensation capacitor. R2 is calculated using the following equation:

$$R2=R1/((V_{OUT}/0.794V)-1)\Omega$$

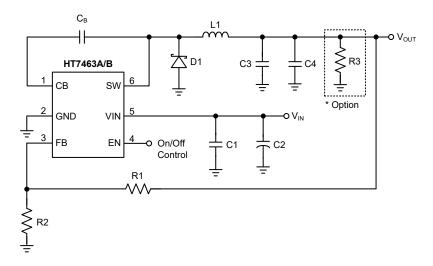
#### **Protection Features**

The devices include dedicated protection circuitry which is fully active during normal operation for full device protection. The thermal shutdown circuitry turns off power to the device when the die temperature reaches excessive levels. The UVLO comparator protects the power device during supply power startup and shutdown to prevent operation at voltages less than the minimum input voltage. The HT7463A/B also features a shutdown mode decreasing the supply current to approximately  $0.1\mu A$ .

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#### **Recommended Component Values**



Note: \* Option: Add a dummy load R3 in order to keep stability when the input voltage is lower than 5.0V and no load. Refer to component selection for low input voltage and no load description below.

#### **Component Recommended Values**

V <sub>OUT</sub> (V)	Package	R1 (kΩ)	R2 (kΩ)
3.3		51 (±1%)	16 (±1%)
5.0	SMD 0603	82 (±1%)	15 (±1%)
12.5		91 (±1%)	6.2 (±1%)

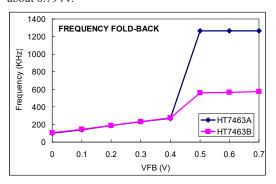
Reference	Package	Description	Part Number	Mfgr.	
C1	SMD 0603	CAP 0.1µF/50V	GRM188R71H104JA93D	Murata	
C2	DIP	68uF/63V	LGK Series	Liket Corp	
C3	SMD 0603	CAP 0.1µF/50V	GRM188R71H104JA93D	Murata	
C4	DIP	47µF/25V	LGK Series	Liket Corp	
СВ	SMD 0603	CAP 0.1µF/50V	GRM188R71M104K9	Murata	
1.1	5.8mm×5.2mm×4.5mm	HT7463A: 15µH/22µH	GS54-150K / GS54-220K	Cong Song	
L'	5.6111111^5.2111111^4.5111111	HT7463B: 33µH/47µH	GS54-330K / GS54-470K	Gang Song	
D1	DO-214AC	Schottky Rectifier	SS16	Fairchild	

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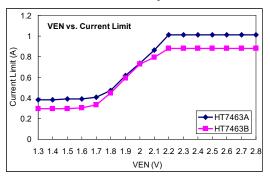
#### **Frequency Fold-back Function**

The devices include a frequency fold-back function to prevent situations of over current when the output is shorted. It efficiently reduces overheating even if the output is shorted. This function is implemented by changing the switching frequency according the feedback voltage, V<sub>FB</sub>. When the output node is shorted, the device will reduce the frequency to 105kHz for the HT7463A/HT7463B respectively resulting in a clamped input current. The HT7463A/HT7463B operates at a frequency of 1250/550kHz under normal conditions and the feedback voltage is about 0.794V.



#### **Start-up Function**

The device EN pin in conjunction with an RC filter is used to tailor the soft-start time to specific application requirements. When a voltage applied to the EN pin is between 0V and 2.3V, the device will cause the cycle-by-cycle current limit in the power stage to be modulated for a minimum current limit at 0V up to a the rated current limit at 2.3V. Thus, the output rise time and inrush current at startup are controlled.



## **Component Selection Guide**

#### Inductor

Use an inductor with a DC current rating at least 25% percent higher than the maximum load current for most applications. The DC resistance of the inductor is a key parameter affecting efficiency. With regard to efficiency, the inductor's DC resistance should be less than  $200 \text{m}\Omega$ . For most application, the inductor value can be calculated from the following equation.

$$L = \frac{V_{out} \times (V_{IN} - V_{out})}{V_{IN} \times I_{ripple} \times f_{sw}}$$

A higher value of ripple current reduces the inductance value, but increases the conductance loss, core loss, and current stress for the inductor and switch devices. A suggested choice is for the inductor ripple current to be 30% of the maximum load current.

If some electromagnetic interference situation, suggest inductor value  $100\mu H$  or more.

#### **Input Capacitor**

A low ESR ceramic capacitor (CIN) is needed between the VIN pin and GND pin. Use ceramic capacitors with X5R or X7R dielectrics for their low ESRs and small temperature coefficients. For most applications, a  $2.2\mu\text{F}$ -  $10\mu\text{F}$  capacitor will suffice.

#### **Output Capacitor**

The selection of COUT is driven by the maximum allowable output voltage ripple. Use ceramic capacitors with X5R or X7R dielectrics for their low ESR characteristics. Capacitors in the range of  $22\mu F$  to  $100\mu F$  are a good starting point with an ESR of  $0.1\Omega$  or less.

 $330\mu F$  is recommended to achieve 1% output ripple requirement.

#### **Schottky Diode**

The breakdown voltage rating of the diode should be higher than the maximum input voltage. The current rating for the diode should be equal to the maximum output current to ensure the best reliability in most applications. In this case it is possible to use a diode with a lower average current rating, however the peak current rating should be higher than the maximum load current.

#### **Bootstrap Capacitor**

A  $0.1\mu F$  ceramic capacitor or larger is recommended for the bootstrap capacitor. Generally a  $0.1\mu F$  to  $1\mu F$  value can be used to ensure sufficient gate drive for the internal switches and a consistently low  $R_{DSON}$ .



# Component Selection for Low-input Voltage & No-load

When the input voltage is lower than 5.0V and no load, in order to keep stability, two methods listed below can be used.

Reduce the low-side feedback resistor to  $0.8 \mathrm{K}\Omega$ . The recommended value is listed below.

V <sub>OUT</sub> (V)	Package	R1 (kΩ)	R2 (kΩ)
3.3	SMD 0603	2.5 (±1%)	0.8 (±1%)

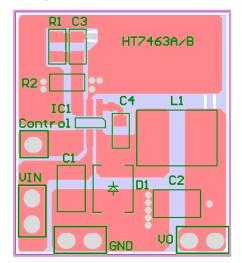
Add a dummy load R3. The value is calculated by the following equation.

$$R3 = V_{OUT}/0.001\Omega$$

# **Layout Consideration Guide**

To reduce problems with conducted noise, there are some important points to consider regarding the PCB layout.

- Ensure all feedback connections are short and direct.
  Place the feedback resistors and compensation components as close to the FB pin as possible.
- The input bypass capacitor must be placed close to the VIN pin.
- The inductor, schottky diode and output capacitor trace should be as short as possible to reduce conducted and radiated noise and increase overall efficiency.
- Keep the power ground connection as short and wide as possible.



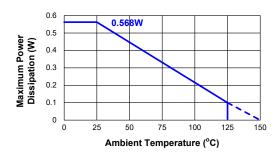
#### **Thermal Considerations**

The maximum power dissipation depends on the thermal resistance of the IC package, the PCB layout, the rate of the surrounding airflow and the difference between the junction and ambient temperature. The maximum power dissipation can be calculated by the following formula:

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

where  $T_{J(MAX)}$  is the maximum junction temperature,  $T_A$  is the ambient temperature and  $q_{JA}$  is the junction-to-ambient thermal resistance of the IC package (220°C/W for 6-pin SOT23)

For maximum operating rating conditions, the maximum junction temperature is 150°C. However, it is recommended that the maximum junction temperature does not exceed 125°C in normal operation to maintain reliability. The derating curve for maximum power dissipation is as follows:





# **Package Information**

Note that the package information provided here is for consultation purposes only. As this information may be updated at regular intervals users are reminded to consult the <u>Holtek website</u> for the latest version of the <u>Package/Carton Information</u>.

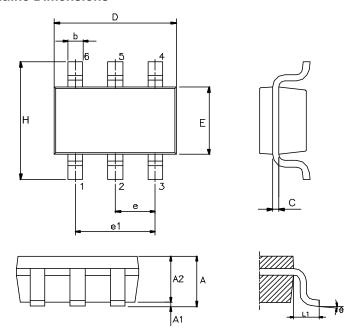
Additional supplementary information with regard to packaging is listed below. Click on the relevant section to be transferred to the relevant website page.

- Package Information (include Outline Dimensions, Product Tape and Reel Specifications)
- The Operation Instruction of Packing Materials
- Carton information

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# 6-pin SOT23 Outline Dimensions



Cymphal	Dimensions in inch			
Symbol	Min.	Nom.	Max.	
A	_	_	0.057	
A1	_	_	0.006	
A2	0.035	0.045	0.051	
b	0.012	_	0.020	
С	0.003	_	0.009	
D	_	0.114 BSC	_	
Е	_	0.063 BSC	_	
е	_	0.037 BSC	_	
e1	_	0.075 BSC	_	
Н	_	0.110 BSC	_	
L1	_	0.024 BSC	_	
θ	0°	_	8°	

Cumbal	Dimensions in mm			
Symbol	Min.	Nom.	Max.	
A	_	_	1.45	
A1	_	_	0.15	
A2	0.90	1.15	1.30	
b	0.30	_	0.50	
С	0.08	_	0.22	
D	_	2.90 BSC	_	
E	_	1.60 BSC	_	
е	_	0.95 BSC	_	
e1	_	1.90 BSC	_	
Н	_	2.80 BSC	_	
L1	_	0.60 BSC	_	
θ	0°	_	8°	

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